

Seminar 1: Difference equations

January 15, 2007

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 - Discrete-time control of continuous-time systems
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Discrete-time control

Digital system

Discrete-time $u(k), y(k)$,
 $k = 0, 1, 2, \dots$

Discrete values a signal can
take (finite precision)

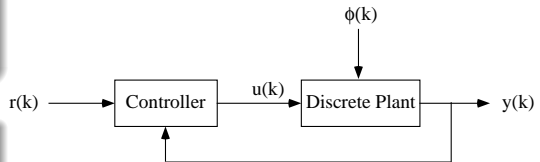
Example

Car engine control

$u(k)$ = fuel injection time

$y(k)$ = air/fuel ratio

k = combustion cycle



Discrete-time control of continuous-time systems

- analogue signals $u(t)$, $y(t)$
- real time $t \in \mathfrak{R}$
- discrete-time signals $u(k)$, $y(k)$
- discrete time $k = 0, 1, 2, \dots$

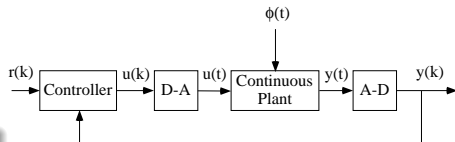
Example

Tank level control

$u(k)$ = pump voltage

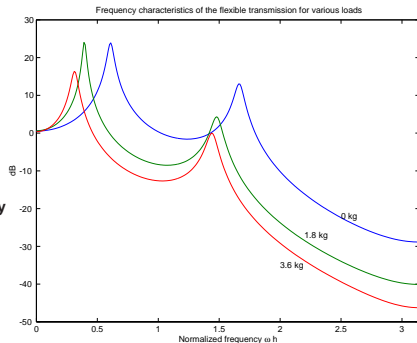
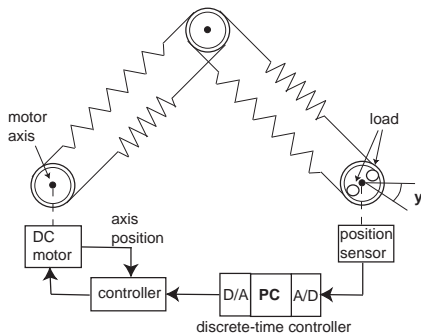
$y(k)$ = tank level

k = index of sampling instant



Discrete-time system includes D-A and A-D converters

A challenging benchmark example



You will learn how to make a model-based controller design handling with changing dynamics

Linear difference equations

A linear difference equation describing a signal y

$$y(k) = -a_1y(k-1) - a_2y(k-2) - \dots - a_ny(k-n)$$

With input u and output y

$$y(k) = -a_1y(k-1) - \dots - a_ny(k-n) + b_0u(k) + b_1u(k-1) + \dots$$

With two inputs, u_1 and u_2 , and one output y

$$y(k) = -a_1y(k-1) - \dots - a_ny(k-n) + b_0u_1(k) + b_1u_1(k-1) + \dots + c_0u_2(k) + c_1u_2(k-1) + \dots$$

where $a_1, a_2, \dots, b_0, b_1, \dots, c_0, c_1, \dots$ are real constants

Example

$$y(k) = -0.9y(k-1) + 0.1u(k-1)$$

Initial conditions $y(0) = u(0) = 0$

Response of the system for the input $u(1) = 1, u(2) = -1$

$$y(1) = -0.9y(0) + 0.1u(0) = -0.9 \cdot 0 + 0.1 \cdot 0 = 0$$

$$y(2) = -0.9y(1) + 0.1u(1) = -0.9 \cdot 0 + 0.1 \cdot 1 = 0.1$$

$$y(3) = -0.9y(2) + 0.1u(2) = -0.9 \cdot 0.1 + 0.1 \cdot (-1) = -0.19$$

Shift operators

Backward-shift operator

$q^{-1}y(k) = y(k-1)$
implementable (memory)

Forward-shift operator

$qy(k) = y(k+1)$
not implementable (future)

Example

$$\begin{aligned}y(k) &= -0.9y(k-1) + 0.1u(k-1) \\y(k) &= -0.9q^{-1}y(k) + 0.1q^{-1}u(k) \\(1 + 0.9q^{-1})y(k) &= 0.1q^{-1}u(k) \\A(q^{-1})y(k) &= B(q^{-1})u(k)\end{aligned}$$

Example

$$\begin{aligned}y(k) &= -0.9y(k-1) + 0.1u(k-1) \\y(k+1) &= -0.9y(k) + 0.1u(k) \\(q + 0.9)y(k) &= 0.1u(k) \\A(q)y(k) &= B(q)u(k)\end{aligned}$$

Polynomial description

A difference equation in **polynomial form**

$$A(q^{-1})y(k) = B(q^{-1})u(k), \quad \begin{cases} A(q^{-1}) = 1 + a_1q^{-1} + a_2q^{-2} + \dots \\ B(q^{-1}) = b_0 + b_1q^{-1} + b_2q^{-2} + \dots \end{cases}$$

or, for brevity,

$$Ay = Bu$$

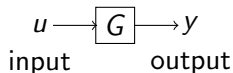
useful for *analysis*.

For *calculation* of responses, use **recursive form**

$$y(k) = -a_1y(k-1) \dots + b_0u(k) + b_1u(k-1) \dots$$

Transfer operator

Input-output description



$$y(k) = G(q^{-1})u(k), \quad G(q^{-1}) = \frac{B(q^{-1})}{A(q^{-1})}$$

Transfer operator G interpreted as difference equation

$$y(k) = -a_1y(k-1) - a_2y(k-2) - \dots + b_0u(k) + b_1u(k-1) \dots$$

Poles and zeros

$$y = G(q^{-1})u = \bar{G}(q)u$$

Forward-shift description

zero

z is a zero if $\bar{G}(z) = 0$

pole

λ is a pole if $\bar{G}(\lambda) \rightarrow \infty$

Example

$$y(k) = 0.9y(k-1) + u(k) + u(k-1)$$
$$\bar{G}(q) = \frac{q+1}{q-0.9} \Rightarrow \begin{cases} \lambda_1 = 0.9 \\ z_1 = -1 \end{cases}$$

Example

Backward-shift description

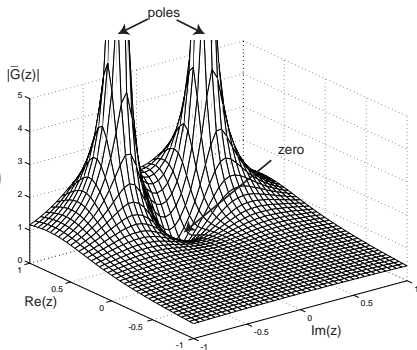
$$A(q^{-1})y(k) = B(q^{-1})u(k)$$

with $\lambda_{1,2} = 0.7 \pm 0.4i$, $z_1 = 0.5$

$$\begin{cases} A(q^{-1}) = (1 - \lambda_1 q^{-1})(1 - \lambda_2 q^{-1}) \\ B(q^{-1}) = q^{-1}(1 - z_1 q^{-1}) \end{cases}$$

Forward-shift description

$$\bar{G} = \frac{q - z_1}{(q - \lambda_1)(q - \lambda_2)}$$



Stability definition

$$y(k) + a_1y(k-1) + \dots + a_ny(k-n) = 0$$

The system is stable if

$$y(k) \rightarrow 0, k \rightarrow \infty$$

for all initial conditions

Example

If $A(q^{-1}) = 1 - \lambda q^{-1}$, the general solution is $y(k) = y(0)\lambda^k$

$$A(q^{-1})y(k) = y(k) - \lambda y(k-1) = y(0)\lambda^k - \lambda y(0)\lambda^{k-1} = 0$$

Stable if and only if $|\lambda| < 1$

Second order system

$$A(q^{-1})y(k) = y(k) + a_1y(k-1) + a_2y(k-2) = 0$$

Factorize in poles

$$A(q^{-1}) = 1 + a_1q^{-1} + a_2q^{-2} = (1 - \lambda_1q^{-1})(1 - \lambda_2q^{-1})$$

General solution

$$y(k) = c_1\lambda_1^k + c_2\lambda_2^k$$

since $A_i = 1 - \lambda_iq^{-1}$ and $y_i(k) = c_i\lambda_i^k$ satisfies $A_iy_i = 0$

$$Ay = A_1A_2(y_1 + y_2) = A_2 \underbrace{[A_1y_1]}_{=0} + A_1 \underbrace{[A_2y_2]}_{=0} = 0$$

Stable if and only if $|\lambda_i| < 1$, $i = 1, 2$

Example

$$y(k) - 2y(k-1) + 0.99y(k-2) = 0$$

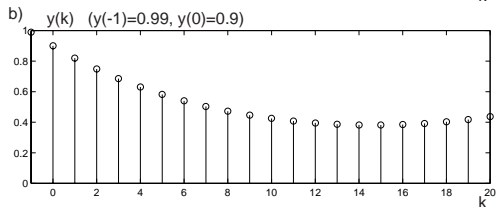
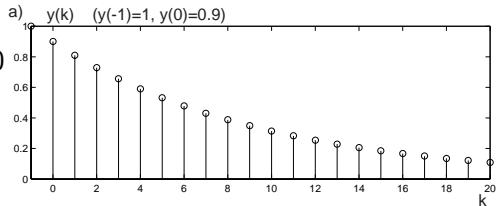
Initial conditions $y(0) = 0.9$
 and

- a) $y(-1) = 1$
- b) $y(-1) = 0.99$

Poles: $\lambda_1 = 0.9$, $\lambda_2 = 1.1$

$$y(k) = c_1 \lambda_1^k + c_2 \lambda_2^k$$

In **a)** $c_2 = 0$, but not in **b)**



Thus, not stable!

General case

Difference equation

$$A(q^{-1})y(k) = 0$$

factorized in poles

$$A(q^{-1}) = (1 - \lambda_1 q^{-1})(1 - \lambda_2 q^{-1}) \dots (1 - \lambda_n q^{-1})$$

Stability criterion

$$|\lambda_i| < 1, \quad i = 1, 2, \dots, n$$

Steady-state (stationary) gain

Unit step

$$u(k) = \begin{cases} 1 & k \geq 0 \\ 0 & k < 0 \end{cases}$$

Step response

$$y(k) + a_1 y(k-1) + \dots + a_n y(k-n) = b_0 \underbrace{u(k)}_{=1} + b_1 \underbrace{u(k-1)}_{=1} + \dots$$

If stable $y(k) \rightarrow y_\infty, k \rightarrow \infty$

$$y_\infty = \frac{b_0 + b_1 + \dots}{1 + a_1 + a_2 + \dots} = \frac{B(1)}{A(1)} = G(1)$$

Steady-state gain

If stable, then $\frac{y_\infty}{u_\infty} = G(1)$

Example: FIR

Unit pulse (impulse)

$$u(k) = \begin{cases} 1 & k = 0 \\ 0 & k \neq 0 \end{cases}$$

Pulse response for $y(k) = 5u(k) + 2u(k - 1)$

$$y(0) = 5u(0) + 2u(-1) = 5 \cdot 1 + 2 \cdot 0 = 5$$

$$y(1) = 5u(1) + 2u(0) = 5 \cdot 0 + 2 \cdot 1 = 2$$

$$y(2) = 5u(2) + 2u(1) = 5 \cdot 0 + 2 \cdot 0 = 0$$

$$y(k) = 0, \quad k > 1$$

Finite Impulse Response (FIR)

Pulse response for $y(k) = b_0u(k) + b_1u(k - 1) + \dots + b_mu(k - m)$

$$y(k) = b_k$$

Example: Step response

Example

Step response for $y(k) = b_0u(k) + b_1u(k-1) + \dots + b_mu(k-m)$

$$y(0) = b_0$$

$$y(1) = b_0 + b_1$$

$$\vdots$$

$$y(m) = b_0 + b_1 + \dots + b_m = B(1)$$

$$y(k) = B(1) \quad k \geq m$$

Example: First order system

Example

Step response for $y(k) = 0.9y(k-1) + 0.1u(k)$

$$y(0) = 0.1$$

$$y(1) = 0.9 \cdot 0.1 + 0.1 = 0.19$$

$$y(2) = 0.9 \cdot 0.19 + 0.1 = 0.271$$

⋮

$$y(k) \rightarrow \frac{B(1)}{A(1)} = 1, \quad k \rightarrow \infty$$

Compare analytical solution

$$y(k) = c_1 \lambda_1^k + 1$$

$$y(0) = c_1 0.9^0 + 1 = 0.1 \rightarrow c_1 = -0.9$$

$$\rightarrow y(k) = -0.9 \cdot 0.9^k + 1$$

Example: First order system

Step response for

$$y(k) = -0.9y(k-1) + 1.9u(k)$$

Recursively

$$y(0) = 1.9$$

$$y(1) = -0.9 \cdot 1.9 + 1.9 = 0.19$$

$$y(2) = -0.9 \cdot 0.19 + 1.9 = 1.73$$

⋮

$$y(k) \rightarrow \frac{B(1)}{A(1)} = 1$$

Analytically

$$y(k) = c_1 \lambda_1^k + 1 = 0.9(-0.9)^k + 1$$

Oscillating response when $\lambda_1 < 0$

Example: Second order system

Pulse response for

$$A(q^{-1})y(k) = B(q^{-1})u(k)$$

$$A(q^{-1}) = (1 - \lambda_1 q^{-1})(1 - \lambda_2 q^{-1})$$

$$k > \deg B \Rightarrow Bu(k) = 0 \text{ (FIR)}$$

$$y(k) = c_1 \lambda_1^k + c_2 \lambda_2^k$$

For complex poles

$$\lambda_1 = \lambda = |\lambda| e^{i \arg(\lambda)}$$

$$\lambda_2 = \lambda^* = |\lambda| e^{-i \arg(\lambda)}$$

$$c_1 = c_2^* = |c| e^{i \arg(c)}$$

$$y(k) = 2|c| |\lambda|^k \cos(\arg(\lambda)k + \arg(c))$$

Oscillating response

amplitude: scales by $|\lambda|^k$

frequency: $\omega = \arg(\lambda)$

Simulation of a difference equation

Implement difference equation

$$y(k) = -a_1 y(k-1) + b_1 u(k-1)$$

Edit the text file:

```

test.sq
functions
{@
function y = mydiff(u)
    global y1 u1 a1 b1
    if isempty(y1),
        y1=0;
        u1=0;
        a1=-0.9;
        b1=0.1;
    end
    y = -a1*y1 + b1*u1;
    y1 = y;
    u1 = u;
}@
    
```

and open it in Sysquake

Test in **command window**

```

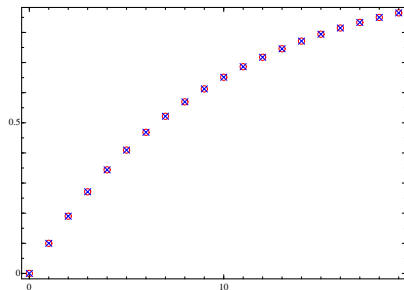
> y=mydiff(0)
y =
    0
> y=mydiff(1)
y =
    0
> y=mydiff(-1)
y =
    0.1000
> y=mydiff(0)
y =
   -0.0100
|
    
```

Example: compare with built-in function

Test in **command window**

```
> global y1 u1 a1 b1  
> y1=0;u1=0;a1=-0.9;b1=0.1;  
> k=0:19;  
> for n=1:20, y(n)=mydiff(1); end;  
> plot(k,y,'ro')  
> A=[1 a1]; B=[0 b1];  
> yf=filter(B,A,ones(1,20));  
> plot(k,yf,'bx')
```

Result in **plot window**

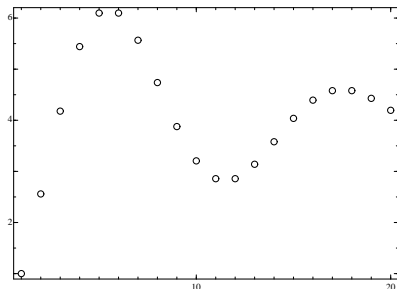


Example: step response of second order system

Test in **command window**

```
> lambda1=0.9*exp(i*pi/6)
lambda1 =
    0.7794 + 0.45j
> lambda2=conj(lambda1)
lambda2 =
    0.7794 - 0.45j
> A=poly([lambda1,lambda2])
A =
     1   -1.5588    0.81
> conv([1 -lambda1],[1 -lambda2])
ans =
     1   -1.5588    0.81
> y=filter(1,A,ones(1,20));
> clf
> plot(y,'o')
> 1/sum(A) % steady-state gain
ans =
    3.9816
```

Result in **plot window**



Notice:

$$\arg \lambda_1 = \frac{\pi}{6} = \omega = \frac{2\pi}{T}$$

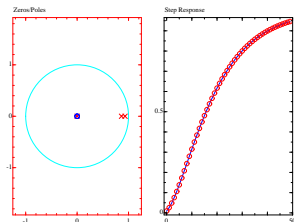
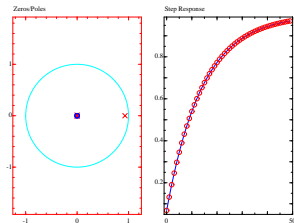
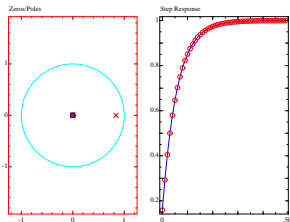
Period $T = 12$ samples

Example: real positive poles

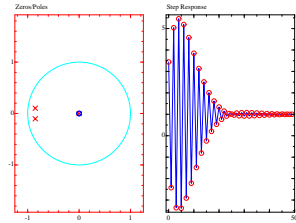
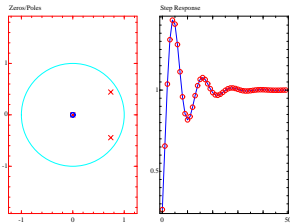
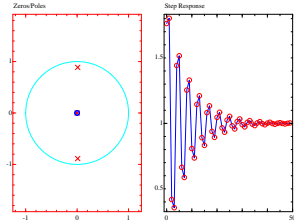
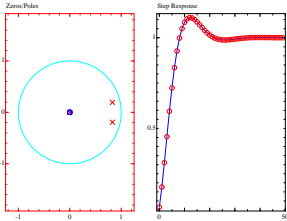
Open **zeropole.sq** and drag poles

$$G(q^{-1}) = \frac{(1 - z_1 q^{-1})(1 - z_2 q^{-1})(1 - z_3 q^{-1})}{(1 - \lambda_1 q^{-1})(1 - \lambda_2 q^{-1})(1 - \lambda_3 q^{-1})}$$

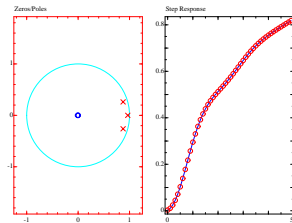
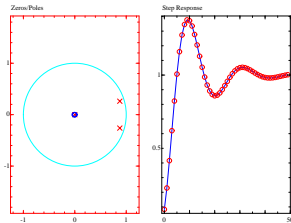
Move λ_1 , then λ_2 (other 0)



Example: complex poles

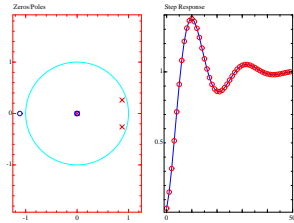
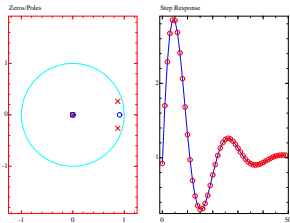
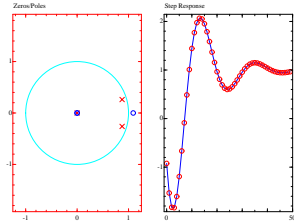
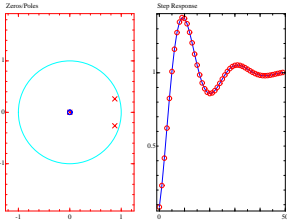


Example: dominant poles



Pole closest to 1 *dominates* step response

Example: influence of zero



Example: influence of zero

Compare step responses (u step)

$$y_1 = G_1 u$$

$$y_2 = G_2 u = G_1 B u$$

$$B = b_0 + b_1 q^{-1} = \frac{1 - zq^{-1}}{1 - z}$$

Study, first step response

$$u_2 = B u \rightarrow \begin{cases} b_0 = \frac{1}{1-z} & k = 0 \\ b_0 + b_1 = 1 & k \geq 1 \end{cases}$$

Now, compare u and u_2 on G_1

$$y_1 = G_1 u$$

$$y_2 = G_1 u_2$$

Scenarios for step responses:

$z = 0.8 \Rightarrow u_2(0) = 5$. Large kick, expected faster and overshoot

$z = -1 \Rightarrow u_2(0) = 0.5$. Smooth step, expected slightly slower

$z = 1.1 \Rightarrow u_2(0) = -10$. Large down-kick, expected undershoot

Conclusion:

\rightarrow zero close to 1 large transient